



COLLEGE OF ENGINEERING AND TECHNOLOGIES
ALMUSTAQBAL UNIVERSITY

Electronics

CTE 207

Lecture 18

- Voltage Divider Bias -

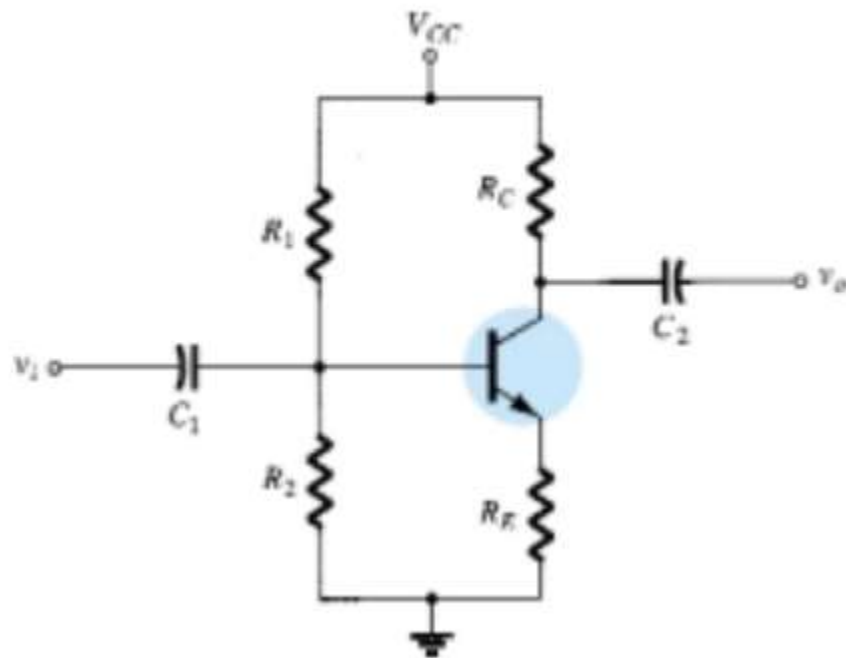
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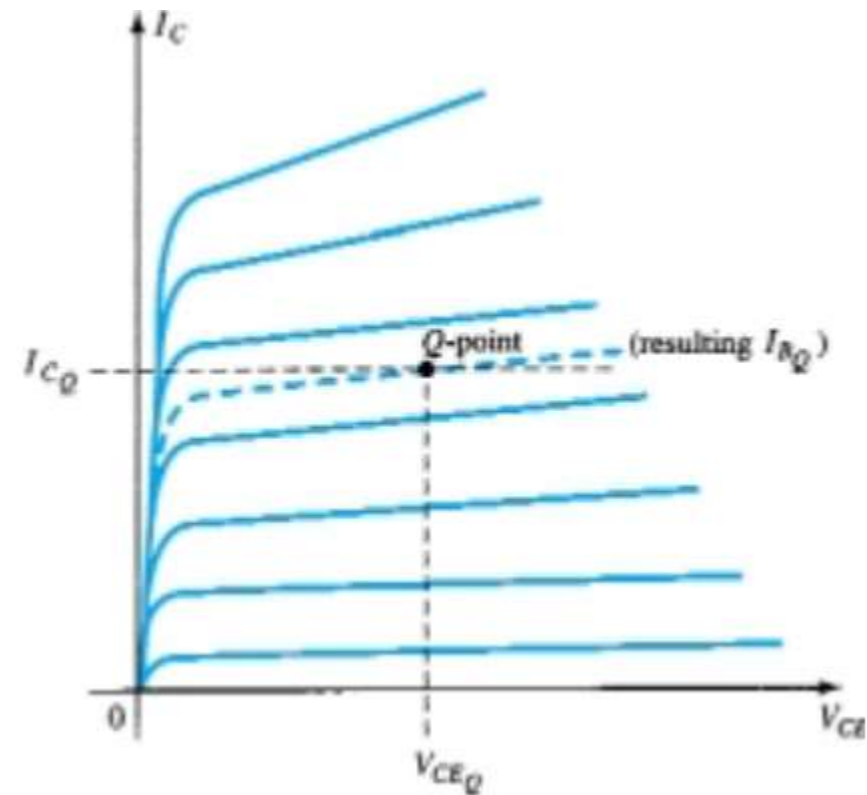
Lecturer / Researcher

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- The voltage-divider bias configuration of Figure below is such a network.
- If analyzed on an exact basis the sensitivity to changes in beta is quite small.
- If the circuit parameters are properly chosen, the resulting levels of I_C and V_{CE} can be almost totally independent of beta.
- There are two methods that can be applied to analyze the voltage divider configuration.

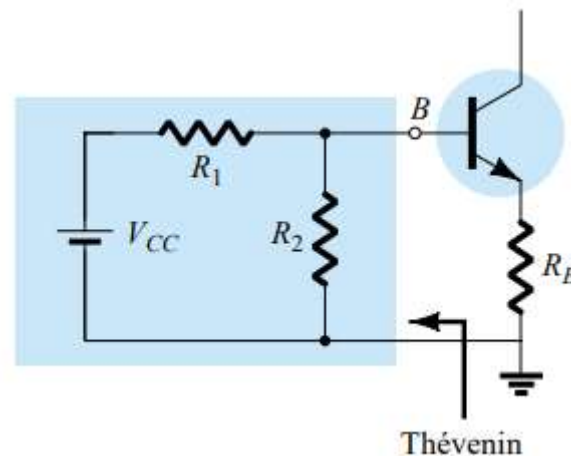


Voltage-divider bias configuration.



Defining the Q-point for the voltage divider bias configuration.

- The input side of the network of Figure above can be redrawn as shown in Figure below for the dc analysis.
- The thevenin equivalent network for the network to the left of the base terminal can then be found in the following manner:



Exact Analysis

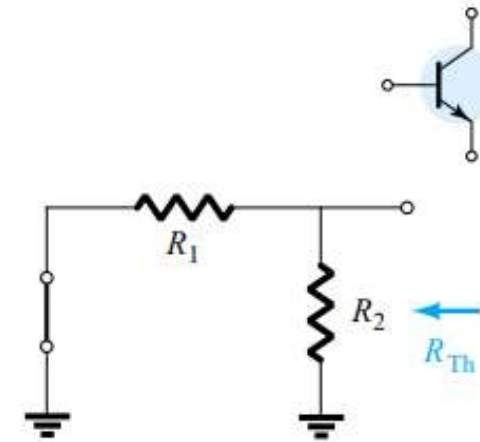
- R_{Th} : The voltage source is replaced by a short-circuit equivalent as shown in Figure below.

$$R_{Th} = R_1 \parallel R_2$$

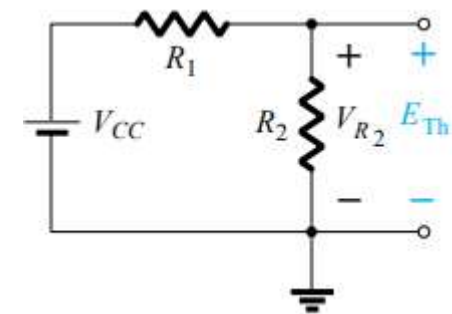
- E_{Th} : The voltage source V_{CC} is returned to the network and the open-circuit Thevenin voltage of Figure below determined as follows:

Applying the voltage-divider rule:

$$E_{Th} = V_{R_2} = \frac{R_2 V_{CC}}{R_1 + R_2}$$



Determining R_{Th}



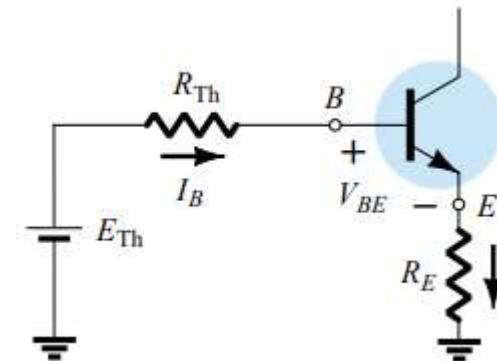
Determining E_{Th}

- The Thevenin network is then redrawn as shown in Figure below, and I_{BQ} can be determined by first applying Kirchhoff's voltage law in the clockwise direction for the loop indicated:

$$E_{Th} - I_B R_{Th} - V_{BE} - I_E R_E = 0$$

Substituting $I_E = (\beta + 1)I_B$ and solving for I_B yields

$$I_B = \frac{E_{Th} - V_{BE}}{R_{Th} + (\beta + 1)R_E}$$



Inserting the Thevenin equivalent circuit

- Once I_B is known, the remaining quantities of the network can be found in the same manner as developed for the emitter-bias configuration.

That is,

$$V_{CE} = V_{CC} - I_C(R_C + R_E)$$

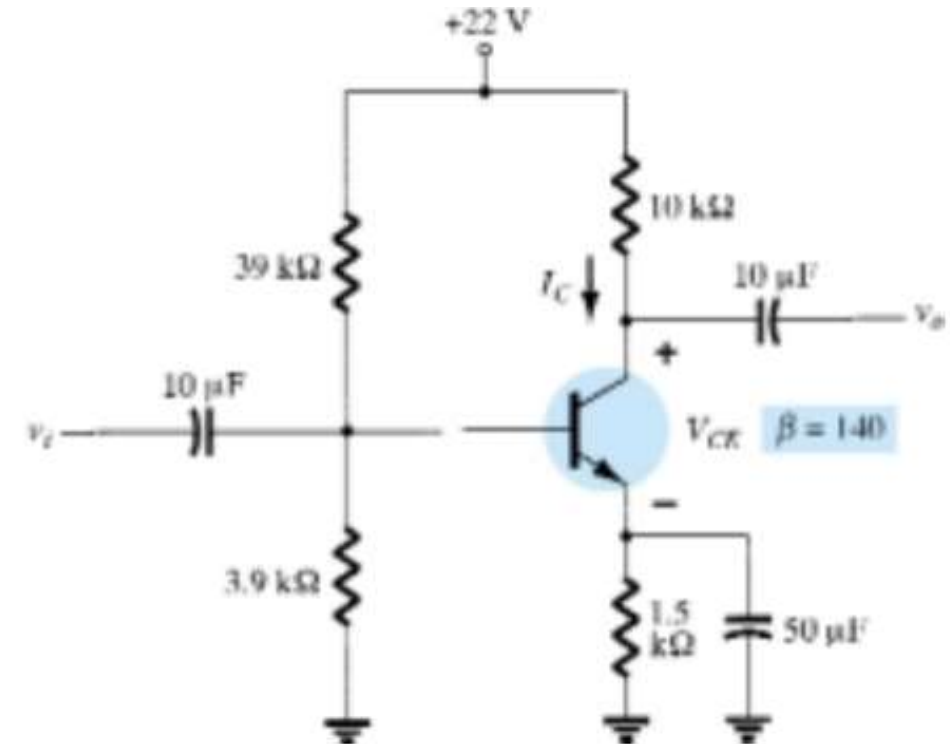
- Which is exactly the same. The remaining equations for V_E , V_C , and V_B are also the same as obtained for the emitter-bias configuration.

Example 1

Determine the DC bias voltage V_{CE} and the current I_C for the voltage-divider configuration of the Figure below.

Sol:

$$\begin{aligned}R_{Th} &= R_1 \parallel R_2 \\ &= \frac{(39 \text{ k}\Omega)(3.9 \text{ k}\Omega)}{39 \text{ k}\Omega + 3.9 \text{ k}\Omega} = 3.55 \text{ k}\Omega \\ E_{Th} &= \frac{R_2 V_{CC}}{R_1 + R_2} \\ &= \frac{(3.9 \text{ k}\Omega)(22 \text{ V})}{39 \text{ k}\Omega + 3.9 \text{ k}\Omega} = 2 \text{ V} \\ I_B &= \frac{E_{Th} - V_{BE}}{R_{Th} + (\beta + 1)R_E} \\ &= \frac{2 \text{ V} - 0.7 \text{ V}}{3.55 \text{ k}\Omega + (141)(1.5 \text{ k}\Omega)} = \frac{1.3 \text{ V}}{3.55 \text{ k}\Omega + 211.5 \text{ k}\Omega} \\ &= 6.05 \mu\text{A}\end{aligned}$$

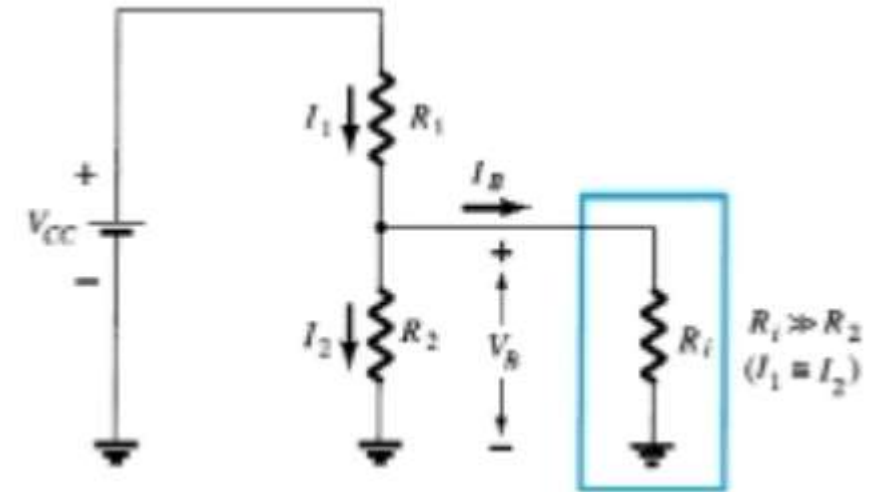


$$\begin{aligned}I_C &= \beta I_B \\ &= (140)(6.05 \mu\text{A}) \\ &= \mathbf{0.85 \text{ mA}}\end{aligned}$$

$$\begin{aligned}V_{CE} &= V_{CC} - I_C(R_C + R_E) \\ &= 22 \text{ V} - (0.85 \text{ mA})(10 \text{ k}\Omega + 1.5 \text{ k}\Omega) \\ &= 22 \text{ V} - 9.78 \text{ V} \\ &= \mathbf{12.22 \text{ V}}\end{aligned}$$

Approximate Analysis

- The input section of the voltage-divider configuration can be represented by the network of the Figure below.
- The resistance R_i is the equivalent resistance between base and ground for the transistor with an emitter resistor R_E .



Partial-bias circuit for calculating the approximate base voltage V_B

The voltage across R_2 , which is actually the base voltage, can be determined using the voltage-divider rule (hence the name for the configuration). That is,

$$V_B = \frac{R_2 V_{CC}}{R_1 + R_2}$$

Since $R_i = (\beta + 1)R_E \cong \beta R_E$ the condition that will define whether the approximate approach can be applied will be the following:

$$\beta R_E \geq 10R_2$$

In other words, if β times the value of R_E is at least 10 times the value of R_2 , the approximate approach can be applied with a high degree of accuracy.

Once V_B is determined, the level of V_E can be calculated from

$$V_E = V_B - V_{BE}$$

Approximate Analysis

and the emitter current can be determined from

$$I_E = \frac{V_E}{R_E}$$

and

$$I_{CQ} \cong I_E$$

The collector-to-emitter voltage is determined by

$$V_{CE} = V_{CC} - I_C R_C - I_E R_E$$

but since $I_E \cong I_C$,

$$V_{CEQ} = V_{CC} - I_C (R_C + R_E)$$

Example 2

Repeat the analysis of Example 1 by using the approximate technique, and compare solutions for I_{CQ} and V_{CEQ}

Sol:

$$\beta R_E \geq 10R_2$$

$$(140)(1.5 \text{ k}\Omega) \geq 10(3.9 \text{ k}\Omega)$$

$$210 \text{ k}\Omega \geq 39 \text{ k}\Omega \text{ (satisfied)}$$

$$\begin{aligned} V_B &= \frac{R_2 V_{CC}}{R_1 + R_2} \\ &= \frac{(3.9 \text{ k}\Omega)(22 \text{ V})}{39 \text{ k}\Omega + 3.9 \text{ k}\Omega} \\ &= 2 \text{ V} \end{aligned}$$

$$\begin{aligned}V_E &= V_B - V_{BE} \\ &= 2 \text{ V} - 0.7 \text{ V} \\ &= 1.3 \text{ V}\end{aligned}$$

$$I_{CQ} \cong I_E = \frac{V_E}{R_E} = \frac{1.3 \text{ V}}{1.5 \text{ k}\Omega} = \mathbf{0.867 \text{ mA}}$$

compared to 0.85 mA with the exact analysis. Finally,

$$\begin{aligned}V_{CEQ} &= V_{CC} - I_C(R_C + R_E) \\ &= 22 \text{ V} - (0.867 \text{ mA})(10 \text{ k}\Omega + 1.5 \text{ k}\Omega) \\ &= 22 \text{ V} - 9.97 \text{ V} \\ &= \mathbf{12.03 \text{ V}}\end{aligned}$$

- The results for I_{CQ} and V_{CEQ} are certainly close, and considering the actual variation in parameter values one can certainly be considered as accurate as the other.
- The larger the level of R_1 compared to R_2 , the closer the approximate to the exact solution.

