

(c) Given



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Let an input $x_2(t)$ produce an output $y_2(t)$. So the differential equation becomes

$$\frac{d^3y_2(t)}{dt^3} + 2\frac{d^2y_2(t)}{dt^2} + 4\frac{dy_2(t)}{dt} + 3y_2^2(t) = x_2(t+1)$$

The linear combination of the above equations becomes

$$a\frac{d^{3}y_{1}(t)}{dt^{3}} + a2\frac{d^{2}y_{1}(t)}{dt^{2}} + a4\frac{dy_{1}(t)}{dt} + a3y_{1}^{2}(t)$$

$$+ b\frac{d^{3}y_{2}(t)}{dt^{3}} + b2\frac{d^{2}y_{2}(t)}{dt^{2}} + b4\frac{dy_{2}(t)}{dt} + b3y_{2}^{2}(t)$$

$$= ax_{1}(t+1) + bx_{2}(t+1)$$

$$= ax_1(t+1) + bx_2(t+1)$$
i.e.
$$\frac{d^3}{dt^3} [ay_1(t) + by_2(t)] + 2\frac{d^2}{dt^2} [ay_1(t) + by_2(t)] + 4\frac{d}{dt} [ay_1(t) + by_2(t)]$$

$$+ 3[ay_1^2(t) + by_2^2(t)] = ax_1(t+1) + bx_2(t+1)$$
Not a weighted sum of outputs

Since one term in LHS is not a weighted sum of outputs, the superposition principle is not valid. Hence the system is non-linear.

- 3. The output depends on future values of input. Hence the system is non-causal.
- 4. All the coefficients of the differential equation are constant. Hence the system is time-invariant.

So the given system is dynamic, non-linear, non-causal and time-invariant.

(b) Given
$$\frac{d^2y(t)}{dt^2} + 2y(t)\frac{dy(t)}{dt} + 3t y(t) = x(t)$$

- 1. The system is described by a differential equation. Hence the system is dynamic.
- 2. There is a term with product of output and its derivative {i.e. y(t)[dy(t)/dt]}. Hence the system is non-linear. This can be proved.

Let an input $x_1(t)$ produce an output $y_1(t)$

Then
$$\frac{d^2y_1(t)}{dt^2} + 2y_1(t)\frac{dy_1(t)}{dt} + 3t y_1(t) = x_1(t)$$

Let an input $x_2(t)$ produce an output $y_2(t)$.

Then
$$\frac{d^2y_2(t)}{dt^2} + 2y_2(t)\frac{dy_2(t)}{dt} + 3ty_2(t) = x_2(t)$$

The linear combination of the above equations gives d^2

$$a\frac{d^{2}y_{1}(t)}{dt^{2}} + a2y_{1}(t)\frac{dy_{1}(t)}{dt} + a3ty_{1}(t) + \frac{bd^{2}y_{2}(t)}{dt^{2}} + b2y_{2}(t)\frac{dy_{2}(t)}{dt^{2}} + b3t y_{2}(t)$$

$$= ax_{1}(t) + bx_{2}(t)$$



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i.e.
$$\frac{d^2}{dt^2} [ay_1(t) + by_2(t)] + 2 \underbrace{\left[ay_1(t) \frac{dy_1(t)}{dt} + by_2(t) \frac{dy_2(t)}{dt} \right]}_{\text{Not a weighted sum of outputs}} + 3t [ay_1(t) + by_2(t)]$$

$$=[ax_1(t) + bx_2(t)]$$

Since one term in LHS is not a weighted sum of outputs, the superposition principle is not valid. Hence the system is non-linear.

- 3. The output depends on present input only. Hence the system is causal.
- 4. All the coefficients of the differential equation are not constants. One coefficient is a function of time. Hence the system is time-variant.

So the given system is dynamic, non-linear, causal and time-variant.

(c) Given
$$y(t) = ev\{x(t)\}$$

$$y(t) = ev\{x(t)\} = \frac{1}{2} [x(t) + x(-t)]$$

1. For positive values of t, the output depends on past values of input and for negative values of t, the output depends on future values of input. Hence the system is dynamic.

2.
$$y(t) = T[x(t)] = \frac{1}{2}[x(t) + x(-t)]$$

For an input $x_1(t)$,

nciple

m is

nce

$$y_1(t) = \frac{1}{2} [x_1(t) + x_1(-t)]$$

For an input $x_2(t)$,

$$y_2(t) = \frac{1}{2} [x_2(t) + x_2(-t)]$$

The weighted sum of outputs is:

$$ay_1(t) + by_2(t) = a\frac{1}{2}[x_1(t) + x_1(-t)] + b\frac{1}{2}[x_2(t) + x_2(-t)]$$
$$= \frac{1}{2}\{[ax_1(t) + bx_2(t)] + [ax_1(-t) + bx_2(-t)]\}$$

The output due to weighted sum of inputs is:

$$y_3(t) = T[ax_1(t) + bx_2(t)] = \frac{1}{2} \left\{ [ax_1(t) + bx_2(t)] + [ax_1(-t) + bx_2(-t)] \right\}$$

$$y_3(t) = ay_1(t) + by_2(t)$$

The weighted sum of outputs is equal to the output due to weighted sum of inputs. Hence superposition principle is valid and the system is linear.



(e)



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 $y(-2) = \frac{1}{2}[x(-2) + x(2)]$

i.e. for negative values of t, the output depends on future values of input. Hence the system is non-causal.

 $y(t) = \frac{1}{2} [x(t) + x(-t)]$ 4. Given

The output due to input delayed by T units is:

e output due to input delayed by
$$T$$
 units is:

$$y(t,T) = T[x(t-T)] = y(t)|_{x(t)=x(t-T)} = \frac{1}{2}[x(t-T) + x(-t-T)]$$

The output delayed by T units is:

ed by
$$T$$
 units is:

$$y(t-T) = y(t)|_{t=t-T} = \frac{1}{2} [x(t-T) + x(-t+T)]$$

 $y(t,T)\neq y(t-T)$

So the system is time-variant.

So the given system is dynamic, linear, non-causal and time-variant.

(d) Given $y(t) = at^2x(t) + bt x(t-4)$

1. The output depends on past inputs. So it requires memory. Hence it is a dynamic system.

 $y(t) = at^2 x(t) + bt x(t-4)$ 2. Given

For an input $x_1(t)$,

$$y_1(t) = at^2 x_1(t) + bt x_1(t-4)$$

For an input $x_2(t)$,

$$y_2(t) = at^2x_2(t) + btx_2(t-4)$$

The weighted sum of outputs is:

$$py_1(t) + qy_2(t) = pat^2x_1(t) + pbt x_1(t-4) + qat^2x_2(t) + qbt x_2(t-4)]$$

$$= at^2[px_1(t) + qx_2(t)] + bt[px_1(t-4) + qx_2(t-4)]$$

The output due to weighted sum of inputs is:

$$y_3(t) = T[px_1(t) + qx_2(t)] = at^2[px_1(t) + qx_2(t)] + bt[px_1(t-4) + qx_2(t-4)]$$

$$y_3(t) = ay_1(t) + by_2(t)$$

Superposition principle is satisfied. Hence the system is linear.

3. The output depends only on the present and past inputs. It does not depend on



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The output due to weighted sum of inputs is:
$$1$$

$$y_3(n) = T[ax_1(n) + bx_2(n)] = [ax_1(n) + bx_2(n)]^2 + \overline{[ax_1(n-1) + bx_2(n-1)]^2}$$

$$y_3(n) \neq ay_1(n) + by_2(n)$$

3. The output does not depend on future values of input. Hence the system is causal

3. The output does not depend on future
$$x$$
.

4. Given $y(n) = T[x(n)] = x^2(n) + \frac{1}{x^2(n-1)}$

The output due to input delayed by k units is:

tput due to input delayed by
$$k = 1$$

$$y(n,k) = T[x(n-k)] = y(n)|_{x(n)=x(n-k)} = x^2(n-k) + \frac{1}{x^2(n-1-k)}$$

The output delayed by k units is:

$$y(n-k) = y(n)|_{n=n-k} = x^2(n-k) + \frac{1}{x^2(n-k-1)}$$

$$y(n, k) = y(n - k)$$

Hence the system is time-invariant.

So the given system is dynamic, non-linear, causal and time-invariant.

Stable and Unstable Systems 2.2.6

A bounded signal is a signal whose magnitude is always a finite value. For example, a sinewave is a bounded signal. A system is said to be bounded-input, bounded-output (BIBO) stable, if and only if every bounded input produces a bounded output. The output of a stable system does not diverge or does not grow unreasonably large.

Let the input signal x(t) be bounded (finite), i.e.

$$|x(t)| \le M_x < \infty$$
 for all t

where M_x is a positive real number.

$$\left|y(t)\right| \leq M_{y} < \infty$$

i.e. if y(t) is also bounded, then the system is BIBO stable. Otherwise, the system is unstable. That is, we say that a system is unstable even if one bounded input produces an unbounded

It is very important to know about the stability of the system. Stability indicates the system. The stability can be found of the system. usefulness of the system. The stability can be found from the impulse response of the system on the impulse response of the system for a point of th which is nothing but the output of the system for a unit impulse response of the system for a unit impulse input. If the impulse response is absolutely integrable for a continuous-time system or absolutely summable for a discrete-





The necessary and sufficient condition for a system to be BIBO stable is given by the BIBO stability criterion

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

where h(t) is the impulse response of the system. This is called BIBO stability criterion. Proof: Consider a linear time-invariant system with x(t) as input and y(t) as output. The $\frac{p_{roof}}{p_{out}}$ and output of the system are related by the convolution integral.

$$y(t) = \int_{-\infty}^{\infty} x(\tau) h(t - \tau) d\tau$$

Taking absolute values on both sides, we have

sides, we have
$$|y(t)| = \left| \int_{-\infty}^{\infty} x(\tau) h(t - \tau) d\tau \right|$$

Using the fact that the absolute value of the integral of the product of two terms is always less than or equal to the integral of the product of their absolute values, we have

$$\left| \int_{-\infty}^{\infty} x(\tau) h(t-\tau) d\tau \right| \leq \int_{-\infty}^{\infty} |x(\tau)| |h(t-\tau)| d\tau$$

If the input $x(\tau)$ is bounded, i.e. there exists a finite number M_x such that,

$$|x(\tau)| \le M_x < \infty$$

$$|x(\tau)| \le M_x < \infty$$

$$|y(t)| \le M_x \int_{-\infty}^{\infty} |h(t - \tau)| d\tau$$

Changing the variables by $m = t - \tau$, the output is bounded if

$$\int_{-\infty}^{\infty} |h(m)| \, dm < \infty$$

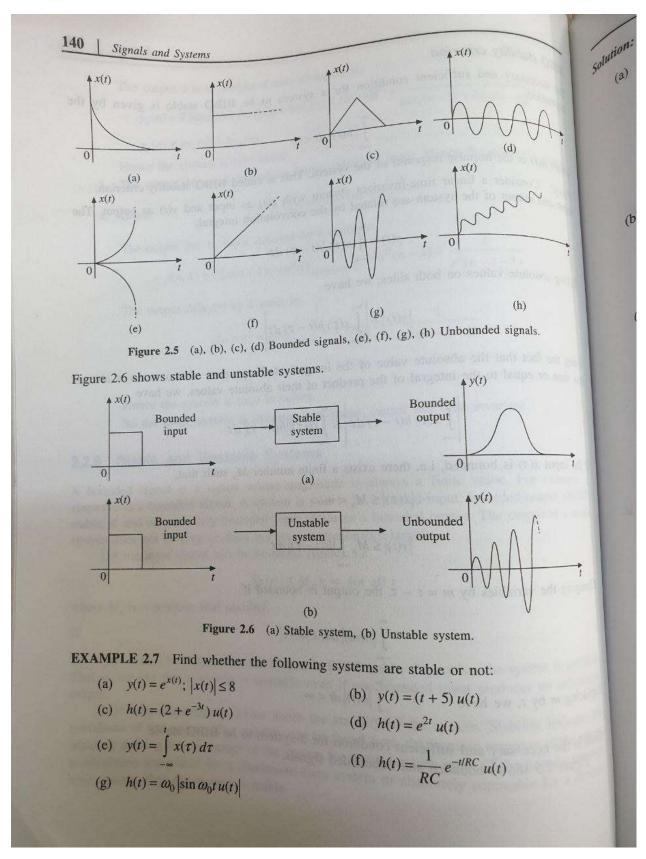
Replacing m by t, we have $\int_{-\infty}^{\infty} |h(t)| dt < \infty$

0)

$$\int\limits_{0}^{\infty}\left|h(t)\right|dt<\infty$$

which is the necessary and sufficient condition for a system to be BIBO stable. Figure 2.5 shows bounded and unbounded signals.







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Solution:

(a) Given

$$y(t) = e^{x(t)}; |x(t)| \le 8$$

Here the input is bounded,

$$|x(t)| \le 8$$

Therefore for stability, the output must be bounded. The output y(t) becomes

$$e^{-8} \le y(t) \le e^8$$

Hence y(t) is also bounded. Therefore, the system is stable. (b) Given y(t) = (t + 5) u(t)

$$y(t) = (t+5) u(t)$$

y(t) = t + 5 for $t \ge 0$

So, as $t \to \infty$, $y(t) \to \infty$ Hence the output grows without any bound and hence the given system is unstable.

(c) Given

$$h(t) = (2 + e^{-3t}) u(t)$$

For a system to be stable,

$$\int_{-\infty}^{\infty} \left| h(t) \right| dt < \infty$$

Here
$$\int_{-\infty}^{\infty} |h(t)| dt = \int_{-\infty}^{\infty} (2 + e^{-3t}) u(t) dt = \int_{0}^{\infty} (2 + e^{-3t}) dt = \left[2t + \frac{e^{-3t}}{-3} \right]_{0}^{\infty} = \infty$$

Since the impulse response is not absolutely integrable, i.e. since $\int |h(t)| dt = \infty$, is not converge and so dupped is not from the system is unstable.

$$h(t) = e^{2t}u(t)$$

$$\int_{-\infty}^{\infty} |h(t)| dt = \int_{-\infty}^{\infty} \left| e^{2t} u(t) \right| dt = \int_{0}^{\infty} e^{2t} dt = \left[\frac{e^{2t}}{2} \right]_{0}^{\infty} = \infty$$

Since the impulse response is not absolutely integrable, i.e. since $\int |h(t)| dt = \infty$, the system is unstable.

(e) Given

$$y(t) = \int_{-\infty}^{t} x(\tau) d\tau$$

For an input $\delta(t)$,

$$y(t) = h(t)$$





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Therefore,

$$h(t) = \int_{-\infty}^{t} \delta(\tau) d\tau = u(t)$$

For stability,

$$\int\limits_{0}^{\infty}\left|h(t)\right|dt<\infty$$

In this case,

$$\int_{-\infty}^{\infty} |h(t)| dt = \int_{-\infty}^{\infty} u(t) dt = \infty$$

Therefore, the impulse response is unbounded and the system is unstable.

 $h(t) = \frac{1}{RC} e^{-t/RC} u(t)$

(f) Given

$$h(t) = \frac{1}{RC}$$

$$\int_{-\infty}^{\infty} |h(t)| dt < \infty$$

For stability,

In this case,

$$\int_{-\infty}^{\infty} |h(t)| dt = \int_{-\infty}^{\infty} \left| \frac{1}{RC} e^{-t/RC} u(t) \right| dt = \int_{0}^{\infty} \left| \frac{1}{RC} e^{-t/RC} \right| dt = 1$$

Therefore, the system is stable.

(g) Given

$$h(t) = \omega_0 \left| \sin \omega_0 t \ u(t) \right|$$

$$\int_{-\infty}^{\infty} |h(\tau)| d\tau = \omega_0 \int_{0}^{\infty} |\sin \omega_0 \tau| d\tau$$

This integral does not converge and so output is not bounded. Hence the system is unstable.

EXAMPLE 2.8 Check the stability of the system defined by

(a)
$$y(n) = ax(n-7)$$

(b)
$$y(n) = x(n) + \frac{1}{2}x(n-1) + \frac{1}{4}x(n-2)$$

(c)
$$h(n) = a^n$$
 for $0 < n < 11$

(d)
$$h(n) = 2^n u(n)$$

(e) h(n) = u(n)

Solution:

(a) Given

$$y(n) = ax(n-7)$$

Let Then

$$x(n) = \delta(n)$$

$$y(n) = h(n)$$

$$h(n) = a\delta(n-7)$$





A system is stable if its impulse response h(n) is absolutely summable.

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

In this case,

$$\sum_{n=-\infty}^{\infty} |h(n)| = \sum_{n=-\infty}^{\infty} a\delta(n-7) = a \quad \text{for } n=7$$

Hence the given system is stable if the value of a is finite.

$$y(n) = x(n) + \frac{1}{2}x(n-1) + \frac{1}{4}x(n-2)$$

Let

$$x(n) = \delta(n)$$

$$(n) = h(n)$$

Then
$$y(n) = h(n)$$

$$h(n) = \delta(n) + \frac{1}{2} \delta(n-1) + \frac{1}{4} \delta(n-2)$$

A discrete-time system is stable if

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

$$(1-\alpha)^{\frac{\alpha}{2}} x d + (\alpha) x d = (\alpha) x$$

The given h(n) has a value only at n = 0, n = 1, and n = 2. For all other values of n from $-\infty$ to ∞ , h(n) = 0.

At
$$n = 0$$
, $h(0) = \delta(0) + \frac{1}{2}\delta(0-1) + \frac{1}{4}\delta(0-2) = \delta(0) + \frac{1}{2}\delta(-1) + \frac{1}{4}\delta(-2) = 1$

At
$$n = 1$$
, $h(1) = \delta(1) + \frac{1}{2}\delta(1-1) + \frac{1}{4}\delta(1-2) = \delta(1) + \frac{1}{2}\delta(0) + \frac{1}{4}\delta(-1) = \frac{1}{2}$

At
$$n = 2$$
, $h(2) = \delta(2) + \frac{1}{2}\delta(2-1) + \frac{1}{4}\delta(2-2) = \delta(2) + \frac{1}{2}\delta(1) + \frac{1}{4}\delta(0) = \frac{1}{4}\delta(1)$

$$\sum_{n=-\infty}^{\infty} |h(n)| = 1 + \frac{1}{2} + \frac{1}{4} = \frac{7}{4} < \infty \text{ (a finite value)}$$

Hence the system is stable.

n is

$$h(n) = a^n$$
 for $0 < n < 11$

$$\sum_{n=-\infty}^{\infty} |h(n)| = \sum_{n=-\infty}^{\infty} |a^n| = \sum_{n=0}^{11} a^n = \frac{1-a^{12}}{1-a}$$

This value is finite for finite value of a. Hence the system is stable if a is finite.





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$$h(n) = 2^n u(n)$$

$$\sum_{n = -\infty}^{\infty} |h(n)| = \sum_{n = -\infty}^{\infty} |2^n u(n)| = \sum_{n = 0}^{\infty} 2^n = \infty$$
les summable. Hence

The impulse response is not absolutely summable. Hence this system is unstable.

(e) Given

$$h(n) = u(n)$$

For stability,

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

In this case,

$$\sum_{n=-\infty}^{\infty} |h(n)| = \sum_{n=0}^{\infty} 1 = 1 + 1 + 1 + \dots = \infty$$

So the output is not bounded and the system is unstable.

EXAMPLE 2.9 Check whether the following digital systems are BIBO stable or not;

- (a) y(n) = ax(n+1) + bx(n-1)
- (b) y(n) = maximum of [x(n), x(n-1), x(n-2)]
- (c) y(n) = ax(n) + b
- (d) $y(n) = e^{-x(n)}$
- (e) $y(n) = ax(n) + bx^2(n-1)$

Solution: Today the roll in a branch and the said to be a sent to be a

(a) Given

$$y(n) = ax(n+1) + bx(n-1)$$

$$x(n) = \delta(n)$$

$$y(n) = h(n)$$

$$y(n) = h(n)$$

Hence the impulse response is:

$$h(n) = a\delta(n+1) + b\delta(n-1)$$

When n = 0, $h(0) = a\delta(1) + b\delta(-1) = 0$

When
$$n = 1$$
, $h(1) = a\delta(2) + b\delta(0) = b$

When
$$n = 2$$
, $h(2) = a\delta(3) + b\delta(1) = 0$

In general,

$$h(n) = \begin{cases} b & \text{for } n = 1 \\ 0 & \text{otherwise} \end{cases}$$

$$\sum_{n=-\infty}^{\infty} |h(n)| = b$$

The necessary and sufficient condition for BIBO stability is:

$$\sum_{n=-\infty}^{\infty} |h(n)| < \infty$$

So the system is BIBO stable if $|b| < \infty$.





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	y(n) = maximum of [x(n), x(n-1), x(n-2)]	(h) Given
	$x(n) = \delta(n)$	(0)
	y(n) = h(n)	If Then
	$h(n) = \text{maximum of } [\delta(n), \delta(n-1), \delta(n-2)]$	The
	$h(0) = \text{maximum of } [\delta(0), \delta(-1), \delta(-2)] = 1$	
	$h(1) = \text{maximum of } [\delta(1), \delta(0), \delta(-1)] = 1$	
	$h(2) = \text{maximum of } [\delta(2), \delta(1), \delta(0)] = 1$	
	$h(3) = \text{maximum of } [\delta(3), \delta(2), \delta(1)] = 0$	
	$h(4) = 0 = h(5) = h(6) \dots$	Similarly
	$\sum_{n=0}^{\infty} h(n) = h(0) + h(1) + h(2) + \cdots$	Ju-
	. n=-∞	2.
	$=1+1+1+0+0+\dots=3$	
	ystem is BIBO stable.	co the giv
	v(n) = ax(n) + b	
	$x(n) = \delta(n)$	(c) Given
	y(n) = h(n)	Then
	ulse response is:	TI-no the
	$h(n) = a\delta(n) + b$	Hence the
	$(0) = a\delta(0) + b = a + b$	When n=
	$(1) = a\delta(1) + b = b$	When n -
		Therefore
	$h(n) = \begin{cases} a+b & \text{when } n=0\\ b & \text{when } n \neq 0 \end{cases}$	Therefore
	$h(n) = \begin{cases} b & \text{when } n \neq 0 \end{cases}$	
is:	and sufficient condition for BIBO stability	
	and sufficient condition	The nece
	$\sum_{n=-\infty}^{\infty} h(n) < \infty$	
Hence, the give	n=-∞	
+···+···	$\sum_{n=-\infty}^{\infty} h(n) = h(0) + h(1) + h(2) + \dots + h(n) $ $= a+b + b + b + \dots + b + \dots$ the ratio between the	
	$\sum h(n) = h(0) + h(1) + h(1) $	Therefore
to torme is on	$= a+b + b + b +\cdots+ b +\cdots$	
successive terms is on	ince the ratio between th	
	converges since the last	This
	ven system is BIBO unstable. $y(n) = e^{-x(n)}$	richee in
	$v(n) = \epsilon$	(d) Given
	$(-\delta(n))$	
	$x(n) = \delta(n)$ $y(n) = h(n)$	If





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Hence the impulse response is:

$$h(n) = e^{-\delta(n)}$$

When
$$n = 0$$
, $h(0) = e^{-\delta(0)} = e^{-1}$

When
$$n = 0$$
, $h(0) = e^{-\delta(1)} = e^{0} = 1$

In general,

$$h(n) = \begin{cases} e^{-1} & \text{when } n = 0\\ 1 & \text{when } n \neq 0 \end{cases}$$

The necessary and sufficient condition for BIBO stability is:

$$\sum_{n=-\infty}^{\infty} \left| h(n) \right| < \infty$$

Therefore,

$$\sum_{n=-\infty}^{\infty} |h(n)| = |h(0)| + |h(1)| + |h(2)| + \dots + |h(n)| + \dots$$

$$= e^{-1} + 1 + 1 + 1 + \dots + 1 + \dots$$

$$= e^{-1} + 1 + 1 + \dots + 1 + \dots$$

$$= e^{-1} + 1 + 1 + \dots + 1 + \dots$$

$$= e^{-1} + 1 + 1 + \dots + 1 + \dots$$

Since the given sequence never converges, it is BIBO unstable.

(e) Given

$$y(n) = ax(n) + bx^2(n-1)$$

If

$$x(n) = \delta(n)$$

Then

$$y(n) = h(n)$$

Hence the impulse response is:

$$h(n) = a\delta(n) + b\delta^{2}(n-1)$$

When
$$n = 0$$
, $h(0) = a\delta(0) + b\delta^{2}(-1) = a$

When
$$n = 1$$
, $h(1) = a\delta(1) + b\delta^{2}(0) = b$

When
$$n = 2$$
, $h(2) = a\delta(2) + b\delta^2(1) = 0$

$$\sum_{n=-\infty}^{\infty} |h(n)| = |h(0)| + |h(1)| + |h(2)| + \dots + |h(n)| + \dots$$
$$= |a| + |b| + 0 + \dots$$

Hence, the given system is BIBO stable if $|a| + |b| < \infty$.

EXAMPLE 2.10 Determine whether each of the system with impulse response/output listed

(a)
$$h(n) = 3^n u(-n)$$

(c)
$$h(n) = \delta(n) + \cos n\pi$$

(e)
$$y(n) = \cos x(n)$$

(b)
$$h(n) = \cos \frac{n\pi}{2}$$

(d)
$$h(n) = e^{3n}u(n-2)$$

(d)
$$h(n) = e^{3n}u(n-2)$$

(f) $y(n) = \sum_{k=-\infty}^{n+5} x(k)$