

1. Voltage-Divider Bias

A more practical bias method is to use V_{cc} as the single bias source, as shown in Figure. To simplify the schematic, the battery symbol is omitted and replaced by a line termination circle with a voltage indicator (V_{cc}) as shown.

A dc bias voltage at the base of the transistor can be developed by a resistive voltage divider that consists of R1 and R2, as shown in Figure.

VCC is the dc collector supply voltage. <u>Two current paths</u> are between point **A** and **ground**: one through **R2** and the other through the **base-emitter junction** of the transistor and RE.

To analyse a voltage-divider circuit in which I_{B} is small compared to I_{2} , first calculate the voltage on the base using the unloaded voltage-divider rule:

$$V_{\rm B} \cong \left(\frac{R_2}{R_1 + R_2}\right) V_{\rm CC}$$

Once you know the base voltage, you can find the voltages and currents in the circuit, as follows:

$$V_{\rm E} = V_{\rm B} - V_{\rm BE}$$

and

$$I_{\rm C} \cong I_{\rm E} = \frac{V_{\rm E}}{R_{\rm E}}$$

Then,

Once you know $V_{\rm C}$ and $V_{\rm E}$, you can determine $V_{\rm CE}$.

 $V_{\rm C} = V_{\rm CC} - I_{\rm C}R_{\rm C}$

$$V_{\rm CE} = V_{\rm C} - V_{\rm E}$$





$$R_{\rm IN(BASE)} = \frac{P_{\rm DC}}{I_{\rm E}}$$



2. Thevenin's Theorem Applied to Voltage-Divider Bias

To analyse a voltage-divider biased transistor circuit for base current loading effects, we will apply Thevenin's theorem to evaluate the circuit.

$$V_{\rm TH} = \left(\frac{R_2}{R_1 + R_2}\right) V_{\rm CC}$$

and the resistance is

$$R_{\rm TH} = \frac{R_1 R_2}{R_1 + R_2}$$



 $V_{\rm TH} - V_{R_{\rm TH}} - V_{\rm BE} - V_{R_{\rm E}} = 0$

(c)

Substituting, using Ohm's law, and solving for V_{TH} ,

(b)

 $V_{\rm TH} = I_{\rm B}R_{\rm TH} + V_{\rm BE} + I_{\rm E}R_{\rm E}$

Substituting $I_{\rm E}/\beta_{\rm DC}$ for $I_{\rm B}$,

 $V_{\rm TH} = I_{\rm E}(R_{\rm E} + R_{\rm TH}/\beta_{\rm DC}) + V_{\rm BE}$

Then solving for $I_{\rm E}$,

 R_2

(a)

$$I_{\rm E} = \frac{V_{\rm TH} - V_{\rm BE}}{R_{\rm E} + R_{\rm TH}/\beta_{\rm DC}}$$

3. Voltage-Divider Biased PNP Transistor

pnp transistor requires bias polarities opposite to the npn. This can be accomplished with a negative collector supply voltage, as in Figure (a), or with a positive emitter supply voltage, as in Figure (b).



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In a schematic, the *pnp* is often drawn upside down so that the supply voltage is at the top of the schematic and ground at the bottom, as in Figure (c).

$$V_{\rm TH} + I_{\rm B}R_{\rm TH} - V_{\rm BE} + I_{\rm E}R_{\rm E} = 0$$

By Thevenin's theorem,

$$V_{\rm TH} = \left(\frac{R_2}{R_1 + R_2}\right) V_{\rm CO}$$
$$R_{\rm TH} = \frac{R_1 R_2}{R_1 + R_2}$$

The base current is

$$I_{\rm B} = \frac{I_{\rm E}}{\beta_{\rm DC}}$$

The equation for I_E is

$$I_{\rm E} = \frac{-V_{\rm TH} + V_{\rm BE}}{R_{\rm E} + R_{\rm TH}/\beta_{\rm DC}}$$

The equation for $I_{\rm E}$ is

$$I_{\rm E} = \frac{V_{\rm TH} + V_{\rm BE} - V_{\rm EE}}{R_{\rm E} + R_{\rm TH}/\beta_{\rm DC}}$$



Solution This circuit has the configuration of Figures 5–14(b) and (c). Apply Thevenin's theorem.

$$V_{\rm TH} = \left(\frac{R_1}{R_1 + R_2}\right) V_{\rm EE} = \left(\frac{22 \,\mathrm{k\Omega}}{22 \,\mathrm{k\Omega} + 10 \,\mathrm{k\Omega}}\right) 10 \,\mathrm{V} = (0.688) 10 \,\mathrm{V} = 6.88 \,\mathrm{V}$$
$$R_{\rm TH} = \frac{R_1 R_2}{R_1 + R_2} = \frac{(22 \,\mathrm{k\Omega})(10 \,\mathrm{k\Omega})}{22 \,\mathrm{k\Omega} + 10 \,\mathrm{k\Omega}} = 6.88 \,\mathrm{k\Omega}$$

 $\begin{cases} R_2 \\ 10 k\Omega \end{cases}$

 $\begin{cases} R_1 \\ 22 k\Omega \end{cases}$

R_E 1.0 kΩ

 $\beta_{DC} = 150$

Use Equation 5–8 to determine $I_{\rm E}$.

$$I_{\rm E} = \frac{V_{\rm TH} + V_{\rm BE} - V_{\rm EE}}{R_{\rm E} + R_{\rm TH}/\beta_{\rm DC}} = \frac{6.88 \text{ V} + 0.7 \text{ V} - 10 \text{ V}}{1.0 \text{ k}\Omega + 45.9 \Omega} = \frac{-2.42 \text{ V}}{1.0459 \text{ k}\Omega} = -2.31 \text{ mA}$$

The negative sign on $I_{\rm E}$ indicates that the assumed current direction in the Kirchhoff's analysis is opposite from the actual current direction. From $I_{\rm E}$, you can determine $I_{\rm C}$ and $V_{\rm EC}$ as follows:

$$I_{\rm C} = I_{\rm E} = 2.31 \text{ mA}$$

$$V_{\rm C} = I_{\rm C}R_{\rm C} = (2.31 \text{ mA})(2.2 \text{ k}\Omega) = 5.08 \text{ V}$$

$$V_{\rm E} = V_{\rm EE} - I_{\rm E}R_{\rm E} = 10 \text{ V} - (2.31 \text{ mA})(1.0 \text{ k}\Omega) = 7.68 \text{ V}$$

$$V_{\rm EC} = V_{\rm E} - V_{\rm C} = 7.68 \text{ V} - 5.08 \text{ V} = 2.6 \text{ V}$$



$$R_{\rm TH} = \frac{R_1 R_2}{R_1 + R_2} = \frac{(68 \,\mathrm{k\Omega})(47 \,\mathrm{k\Omega})}{(68 \,\mathrm{k\Omega} + 47 \,\mathrm{k\Omega})} = 27.8 \,\mathrm{k\Omega}$$

Use Equation 5–7 to determine
$$I_{\rm E}$$
.

$$I_{\rm E} = \frac{-V_{\rm TH} + V_{\rm BE}}{R_{\rm E} + R_{\rm TH}/\beta_{\rm DC}} = \frac{2.45 \text{ V} + 0.7 \text{ V}}{2.2 \text{ k}\Omega + 371 \Omega}$$
$$= \frac{3.15 \text{ V}}{2.57 \text{ k}\Omega} = 1.23 \text{ mA}$$

From $I_{\rm E}$, you can determine $I_{\rm C}$ and $V_{\rm CE}$ as follows:

$$I_{\rm C} = I_{\rm E} = 1.23 \text{ mA}$$

$$V_{\rm C} = -V_{\rm CC} + I_{\rm C}R_{\rm C} = -6 \text{ V} + (1.23 \text{ mA})(1.8 \text{ k}\Omega) = -3.79 \text{ V}$$

$$V_{\rm E} = -I_{\rm E}R_{\rm E} = -(1.23 \text{ mA})(2.2 \text{ k}\Omega) = -2.71 \text{ V}$$

$$V_{\rm CE} = V_{\rm C} - V_{\rm E} = -3.79 \text{ V} + 2.71 \text{ V} = -1.08 \text{ V}$$